ATS601LSG

Single Element, Tooth Detecting Speed Sensor IC

Features and Benefits

- Digital output representing target profile
- Single Hall element for twist–insensitive mounting and direction insensitive polarity
- True zero–speed operation
- Defined power-on state
- Air gap independent switchpoints
- Chopper stabilized
- Operation down to 3.3 V
- Factory programmable options:
  - Output polarity: low or high opposite tooth
  - Startup hysteresis: 40 G or 80 G
  - Threshold location: 30% or 50% from signal peak
  - Threshold update: single tooth (continuous) or four–tooth memory
- Fully synchronous digital logic with Scan Path and IDDQ testing

Package: 4–pin SIP (suffix SG)



Not to scale

Description

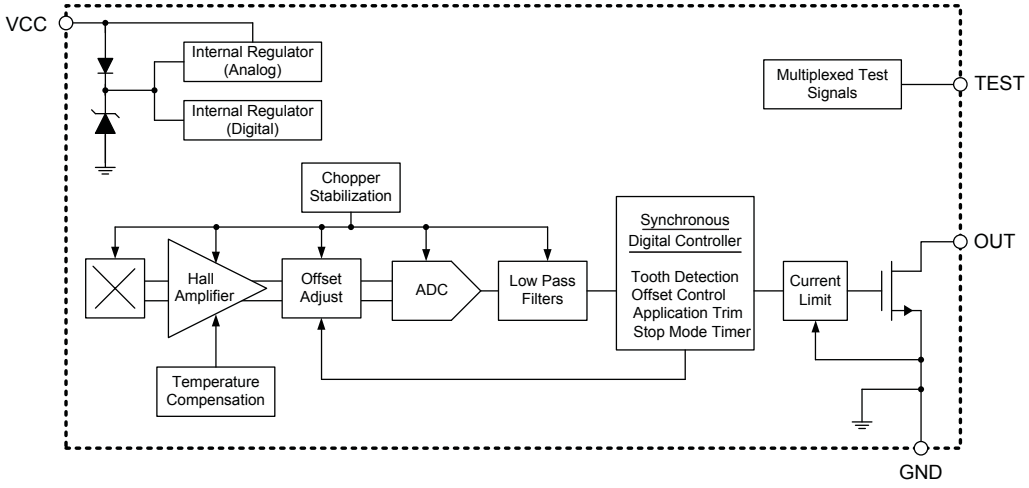
The ATS601LSG is a unique addition to the Allegro™ camshaft sensor IC family of products. As a single element, defined power–on state, sensor IC, it provides first falling edge detection, high running mode edge accuracy, and direction/orientation insensitivity over the full operating range of air gap, speed, and temperature. The ATS601LSG consists of an optimized Hall IC and a simple magnetic pellet configuration.

The single element Hall IC and magnetic pellet configuration switches in response to magnetic signals created by a ferromagnetic target. The IC contains a low bandwidth filter that increases the noise immunity and the signal–to–noise ratio of the sensor IC. Sophisticated digital circuit design eliminates the detrimental effects of target geometry discontinuity, magnet and system offsets, and thermal gradients. Signal processing is used to provide zero speed performance independent of air gap and also to dynamically adapt device performance to the typical operating conditions found in automotive applications, particularly camshaft sensing. The resulting output of the device is a digital representation of the ferromagnetic target profile.

A number of factory programmable options allow for performance optimization to meet specific application requirements.

The ATS601 is provided in a 4–pin SIP package (suffix SG) that is lead (Pb) free, with 100% matte tin leadframe plating.

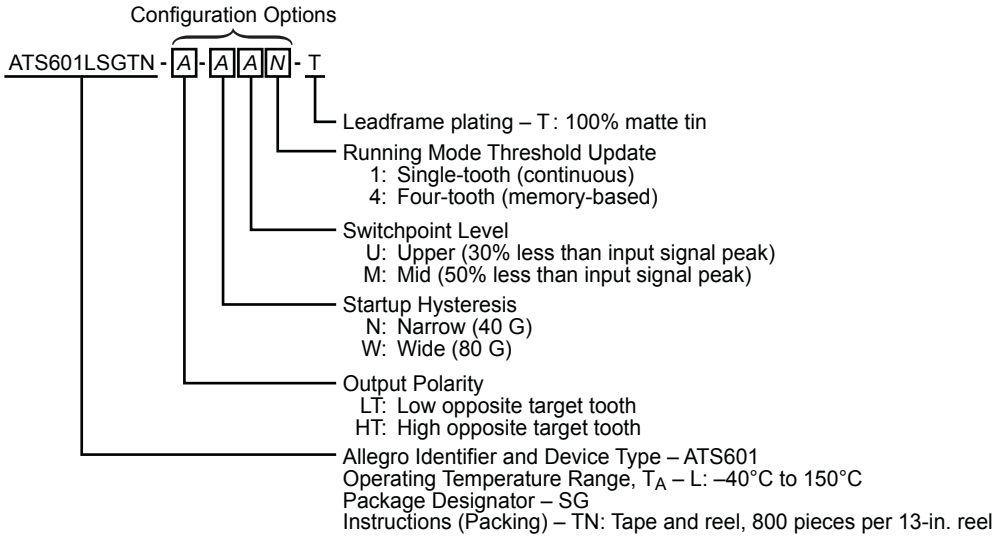
Functional Block Diagram



Selection Guide

Part Number	Packing*
ATS601LSGTN–[*A*]–[*AAN*]–T	800 pieces per 13–in. reel

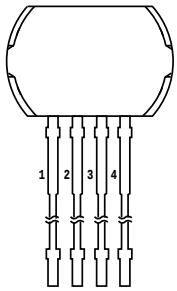
*Contact Allegro™ for additional packing options.



Absolute Maximum Ratings

Characteristic	Symbol	Notes	Rating	Unit
Supply Voltage	V_{CC}		27	V
Reverse Supply Voltage	V_{RCC}		–18	V
Reverse Supply Current	I_{RCC}		–50	mA
Output Voltage	V_{OUT}		27	V
Reverse Output Voltage	V_{ROUT}	R_{PU} > 1000 Ω	–0.5	V
Output Current	I_{OUT}	Internal current limiting is intended to protect the device from output short circuits, but is not intended for continuous operation.	25	mA
Reverse Output Current	I_{ROUT}	V_{OUT} > –0.5 V, T_A = 25°C	–50	mA
Operating Ambient Temperature	T_A	L temperature range	–40 to 150	ºC
Maximum Junction Temperature	T_J(max)		165	ºC
Storage Temperature	T_{stg}		–65 to 170	ºC

Pin–out Diagram



Terminal List Table

Name	Number	Function
VCC	1	Supply voltage
OUT	2	Open drain output
TEST	3	Test pin (MUX)
GND	4	Ground

OPERATING CHARACTERISTICS Valid using Allegro reference target 8X, and V_{CC} and T_A within specification; unless otherwise specified

Characteristic	Symbol	Test Conditions		Min.	Typ.[1]	Max.	Unit[2]
Electrical Characteristics							
Supply Voltage[3]	V_{CC}	Continuous operation, $T_J < T_J$(max)		3.3	–	24	V
Undervoltage Lockout[4]	$V_{CC(UV)rise}$	Rising V_{CC} (0 V → 5 V)		2.5	–	3.1	V
	$V_{CC(UV)fall}$	Falling V_{CC} (5 V → 0 V)		2.4	–	3.0	V
Supply Zener Clamp Voltage	$V_{Zsupply}$	$I_{CC} = I_{CC}$(max) + 3 mA; T_A = 25°C		27	–	–	V
Reverse Supply Zener Clamp Voltage	$V_{RZsupply}$	I_{CC} = –3 mA, T_A = 25°C		–18	–	–	V
Supply Current	I_{CC}			4	–	10	mA
Power-On Characteristics							
Power-On Time[5]	t_{PO}	$V_{CC} > V_{CC}$(min)		–	–	1	ms
Power-On State[6]	POS	Connected as in figure 1		–	High	–	V
Output Stage Characteristics							
Output On Voltage	$V_{OUT(SAT)}$	I_{OUT} = 15 mA, Output = On state (V_{OUT} = Low)		–	–	450	mV
Output Zener Clamp Voltage	$V_{Zoutput}$	I_{OUT} = 3 mA, T_A = 25°C		27	–	–	V
Output Current Limit	$I_{OUT(LIM)}$	Output = On state (V_{OUT} = Low)		25	–	80	mA
Output Leakage Current	$I_{OUT(OFF)}$	V_{OUT} = 24 V, Output = Off state (V_{OUT} = High)		–	–	10	µA
Output Rise Time	t_r	Measured 10% to 90% of V_{OUT}; R_{PU} = 1 kΩ, C_L = 4.7 nF, V_{PU} = 5 V; see figure 2		–	10	–	µs
Output Fall Time	t_f	Measured 90% to 10% of V_{OUT}; R_{PU} = 1 kΩ, C_L = 4.7 nF, V_{PU} = 5 V; see figure 2		3	6	10	µs
Output Delay Time[7]	t_d	1 kHz sinusoidal input signal; see figure 3		–	50	–	µs
Performance Characteristics							
Operational Air Gap Range[8]	AG	N option	Allegro 8X reference target	1.0	–	3.0	mm
		W option		0.5	–	2.5	mm
Magnetic Signal Range	B_{SIG}	N option	Peak to peak signal	60	–	1000	G_{PKPK}
		W option		120	–	1000	G_{PKPK}
Analog Signal Bandwidth	BW	Equivalent to –3 dB cutoff frequency		–	8	–	kHz

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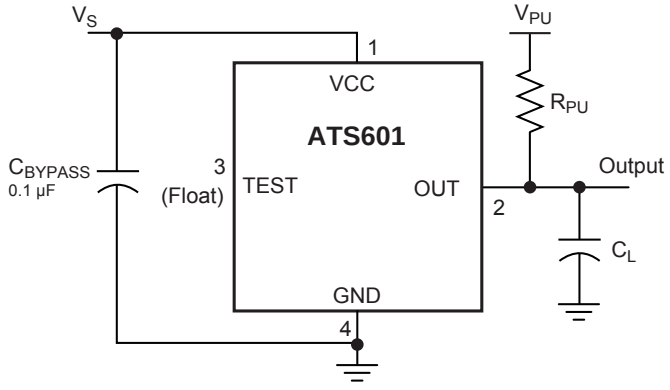


Figure 1. Typical application circuit

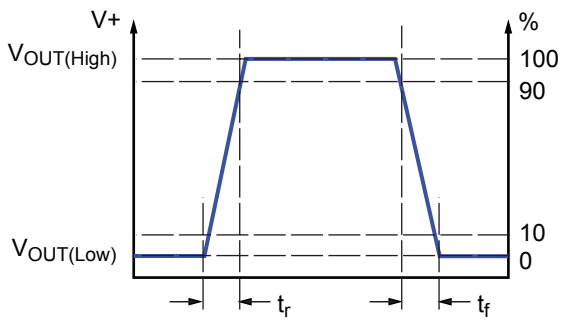


Figure 2. Output Rise Time and Output Fall Time

OPERATING CHARACTERISTICS (continued) Valid using Allegro reference target 8X, and V_{CC} and T_A within specification; unless otherwise specified

Characteristic	Symbol	Test Conditions		Min.	Typ.[1]	Max.	Unit[2]
Operating Mode Characteristics							
Output Polarity	V_{OUT}	LT option	Opposite target tooth, connected as in figure 1	–	Low	–	V
		HT option		–	High	–	V
Startup Hysteresis	$B_{HYS(su)}$	N option	Amount of signal movement needed to generate first output edge	–	40	–	G
		W option		–	80	–	G
Threshold Update Memory	–	1 option	Number of target teeth (peaks) stored in memory for threshold update algorithm	–	1	–	tooth
		4 option		–	4	–	tooth
Running Mode Switchpoint	B_{ST}	U option	% of peak-to-peak, referenced to tooth signal, see figure 4	25	30	35	%
		M option		45	50	55	%
Running Mode Internal Hysteresis	$B_{HYS(rm)}$	% of peak-to-peak signal		5	10	15	%
Stop Mode Timer Period	t_{SM}	Timer interval to initiate Stop mode, no sensed magnetic edges		–	5	–	s
Chopper Frequency	f_C			–	167	–	kHz

[1]Typical values are at T_A = 25°C and V_{CC} = 12 V. Performance may vary for individual units, within the specified maximum and minimum limits.
[2]1 G (gauss) = 0.1 mT (millitesla).
[3]Maximum voltage must be adjusted for power dissipation and junction temperature; see Power Derating section.
[4]Between V_{CC}(min) and $V_{CC(UV)}$ output switching continues to occur but device performance is not guaranteed.
[5]Power–On Time consists of the time from when V_{CC} rises above V_{CC}(min) until the earliest output edge is possible.
[6]Independent of output polarity option (HT or LT).
[7]Output Delay Time is the duration from when a crossing of the magnetic signal switchpoint, B_{ST}, occurs to when the electrical output signal, V_{OUT}, reaches 90% of V_{OUT}(high).
[8]Reduced minimum air gap with N option due to potential for extra switching on first tooth due to magnetic overshoot larger than startup hysteresis, characteristic of the Allegro 8X reference target. Device functions properly in Running mode down to 0.5 mm air gap.

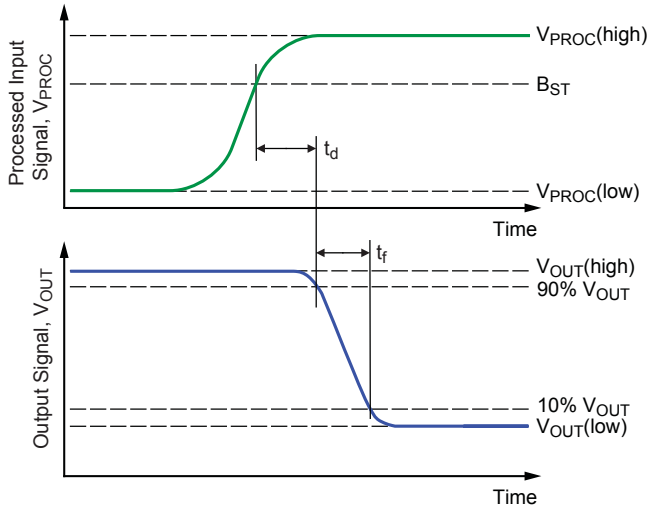


Figure 3. Output Delay Time and Output Fall Time

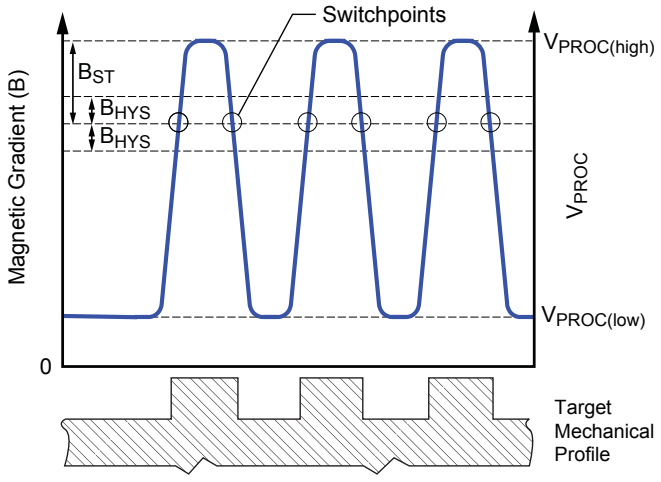


Figure 4. Running Mode Switchpoint and Internal Hysteresis

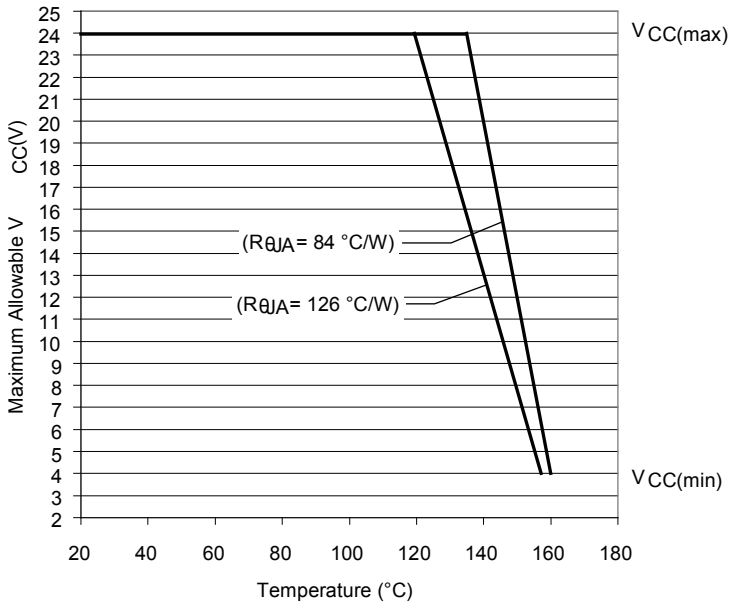
Power Derating

Thermal Characteristics may require derating at maximum conditions, see Power Derating section

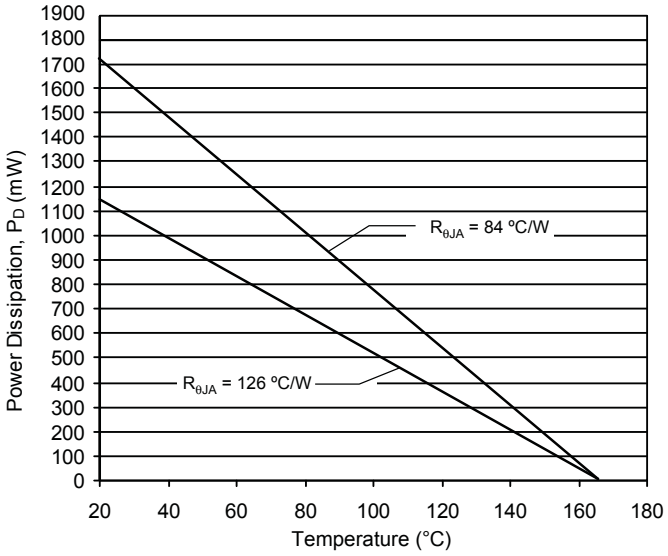
Characteristic	Symbol	Test Conditions*	Value	Unit
Package Thermal Resistance	$R_{\theta JA}$	Single layer PCB, with copper limited to solder pads	126	°C/W
		Single layer PCB, with copper limited to solder pads and 3.57 in.^{2} (23.03 cm^2) copper area each side	84	°C/W

*Additional thermal information available on the Allegro website

Power Derating Curve

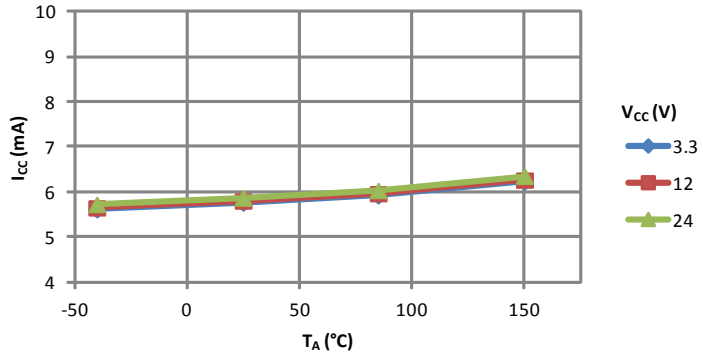


Power Dissipation versus Ambient Temperature

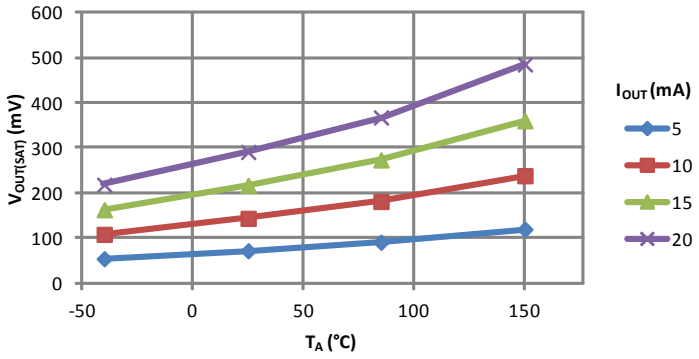


Characteristic Performance

Supply Current versus Ambient Temperature

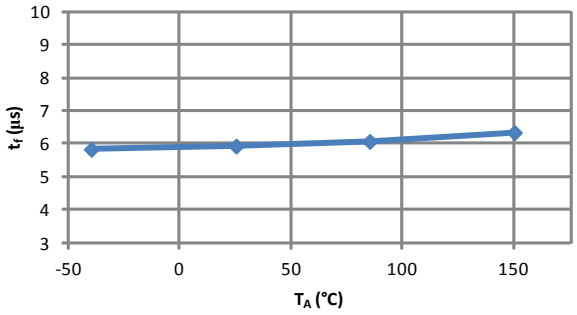


Output On Voltage versus Ambient Temperature

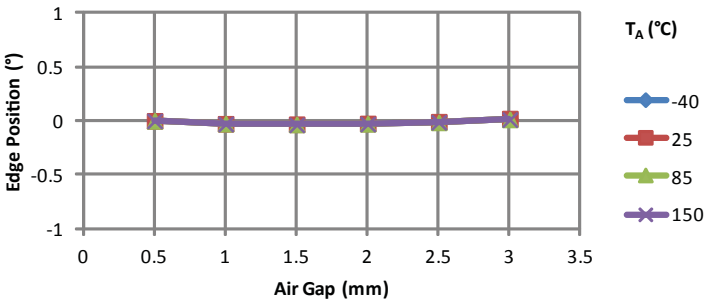


Output Fall Time versus Ambient Temperature

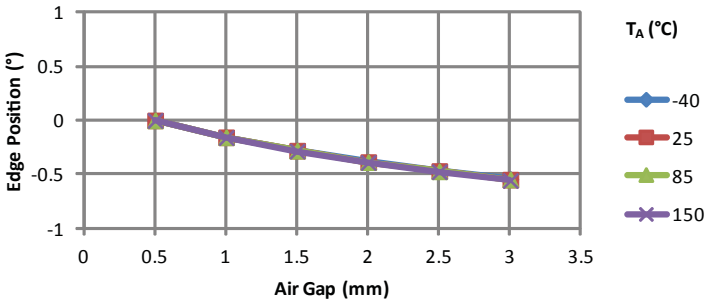
V_{PU} = 5V, R_{PU} = 1kΩ, C_L = 4.7 nF



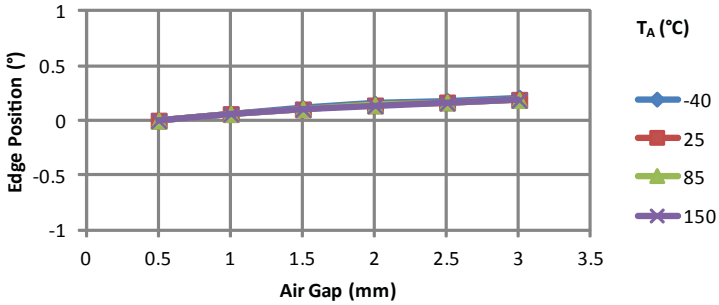
Relative Timing Accuracy versus Air Gap
Rising Mechanical Edge, "U" Switchpoint Option
1000 rpm, Relative to 0.5 mm Air Gap



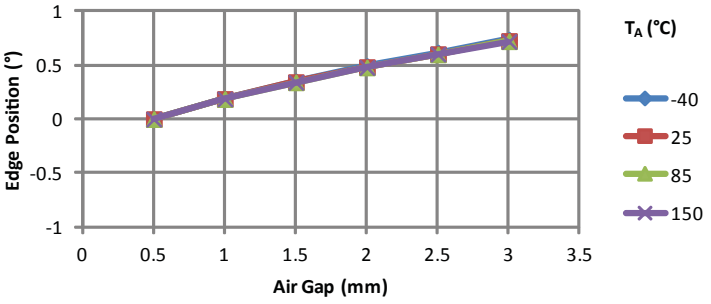
Relative Timing Accuracy versus Air Gap
Rising Mechanical Edge, "M" Switchpoint Option
1000 rpm, Relative to 0.5 mm Air Gap



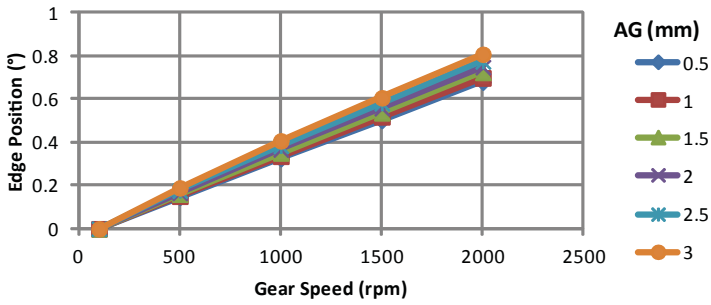
Relative Timing Accuracy versus Air Gap
Falling Mechanical Edge, "U" Switchpoint Option
1000 rpm, Relative to 0.5 mm Air Gap



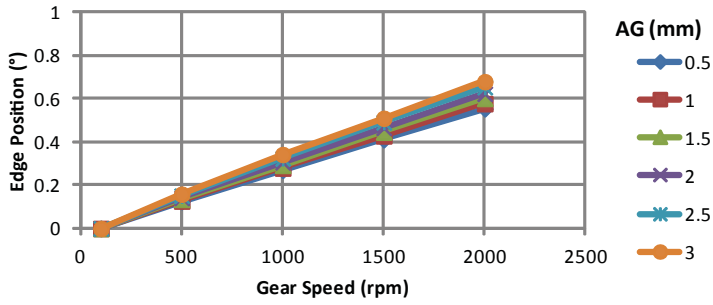
Relative Timing Accuracy versus Air Gap
Falling Mechanical Edge, "M" Switchpoint Option
1000 rpm, Relative to 0.5 mm Air Gap



Relative Timing Accuracy versus Speed
Rising Electrical Edge
25°C, Relative to 100 rpm Gear Speed

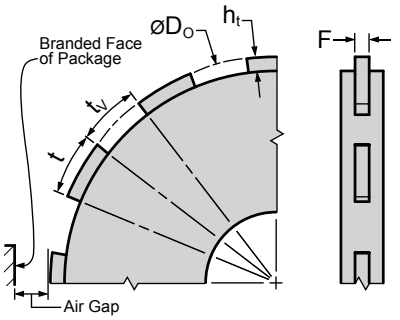


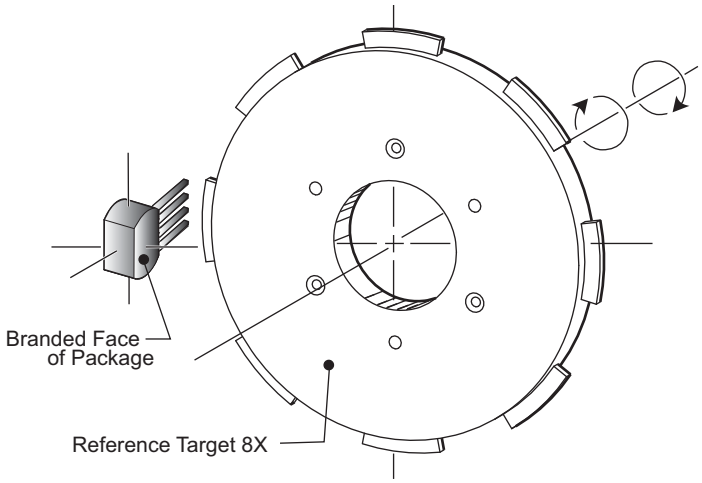
Relative Timing Accuracy versus Speed
Falling Electrical Edge
25°C, Relative to 100 rpm Gear Speed



Reference Target Characteristics

REFERENCE TARGET 8X

Characteristic	Symbol	Test Conditions	Typ.	Units	Symbol Key
Outside Diameter	D_o	Outside diameter of target	120	mm	
Face Width	F	Breadth of tooth, with respect to branded face	6	mm	
Circular Tooth Length	t	Length of tooth, with respect to branded face; measured at D_o	23.6	mm	
Circular Valley Length	t_v	Length of valley, with respect to branded face; measured at D_o	23.6	mm	
Tooth Whole Depth	h_t		5	mm	
Material		CRS 1018	–	–	



Functional Description

Sensing Technology

The ATS601LSG contains a single-chip Hall effect sensor IC, a 4-pin leadframe and a specially designed permanent magnet. The Hall IC supports a chopper stabilized Hall element that measures the magnetic gradient created by the passing of a ferromagnetic object. The difference in the magnetic field strength created by teeth and valleys allows the device to generate a digital output signal that is representative of the target features, independent of the direction of target rotation or rotational orientation of the sensor IC.

The ATS601LSG gear tooth sensor device contains a self-calibrating Hall effect IC that provides a Hall element, a temperature compensated amplifier, and offset cancellation circuitry. The IC also contains a voltage regulator that provides supply noise rejection over the operating voltage range. The Hall transducers and the electronics are integrated on the same silicon substrate by a proprietary BiCMOS process. Changes in temperature do not greatly affect this device due to the stable amplifier design and the offset rejection circuitry.

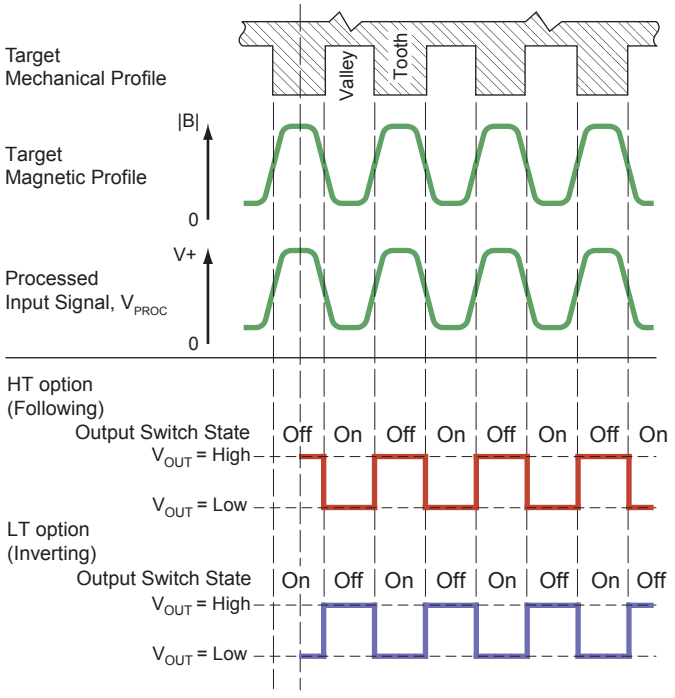


Figure 5. Output Polarity (when connected as shown in figure 1)

Output Polarity

The polarity of the output is programmable to either be low opposite target teeth (LT option) or high opposite target teeth (HT option). See figure 5.

Start-Up Detection

The ATS601LSG always powers-on in the high output state, independent of the polarity option (HT or LT) selected. The output will transition to the low state at the first mechanical edge corresponding to a high-to-low output transition. See figure 6.

Threshold Update

The ATS601 has two programmable options for the threshold update used to establish running mode switching levels. With single-tooth update (1 option) the switching threshold for a tooth

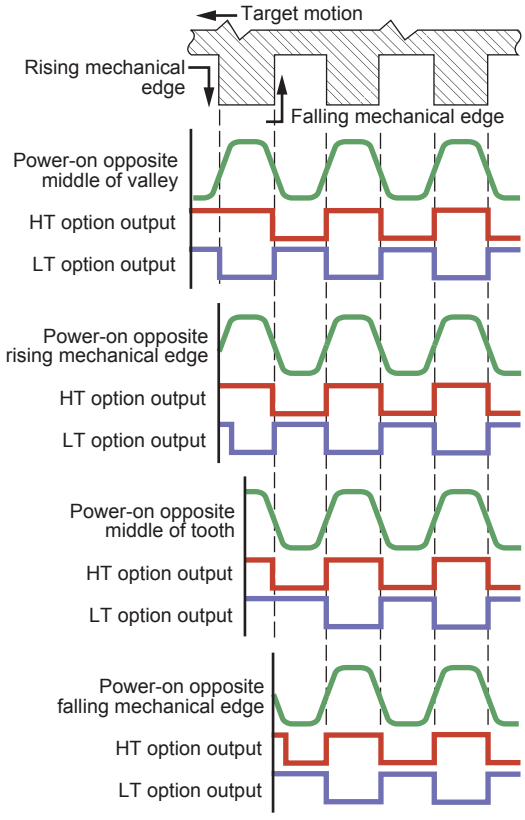


Figure 6. Start-up behavior (when connected as shown in figure 1)

is established based on the measured peak value of the previous tooth. This option can be used with any number of teeth-targets, and is comparable to the Continuous Update mode used on many Allegro sensor ICs.

With the four-tooth update (4 option), peak information from the last four teeth is stored in on-chip memory. Switching thresholds for the upcoming tooth are established based on the stored information from four teeth earlier. When used with four-tooth targets, this is allows for optimized switchpoints based on the same tooth from the previous revolution of the target. The programmable threshold update results in improved output switching accuracy on targets with runout and tooth-to-tooth variation (including narrow valleys). With the four-tooth update option, during the first target rotation the device behaves the same as in single-tooth update while one rotation of target information is stored to memory. Similarly, if a direction change or other sudden magnetic signal change is detected, the device returns to single-tooth update mode and re-learns four target teeth before returning to four-tooth update mode.

Switchpoints

The running mode switchpoints in the ATS601LSG are established dynamically as a percentage of the amplitude of the signal, V_{PROC}. There are two switchpoint (B_{ST}) choices, selectable with the U or M options. The ATS601LSG uses a single switching threshold (operate and release point identical) with internal hysteresis.

The internal running mode hysteresis ($B_{HYS(rm)}$) allows for high performance switching accuracy on both rising and falling edges while maintaining immunity to false switching on noise, vibration, backlash, or other transient events. Figure 7 demonstrates the function of this hysteresis when switching on an anomalous peak.

Power Supply Protection

The ATS601LSG contains an on-chip regulator and can operate over a wide range of supply voltage levels. For applications using an unregulated power supply, transient protection may be added externally. For applications using a regulated supply line, EMI and RFI protection may still be required. Contact Allegro for information on EMC specification compliance.

When the supply voltage falls below the undervoltage lockout level, $V_{CC(UV)fall}$, the device switches to the Off (V_{OUT} = high) state. The device remains in that state until the voltage level is restored to the V_{CC} operating range. Changes in the target magnetic profile have no effect until voltage is restored. This prevents false signals caused by undervoltage conditions from propagating to the output of the sensor IC.

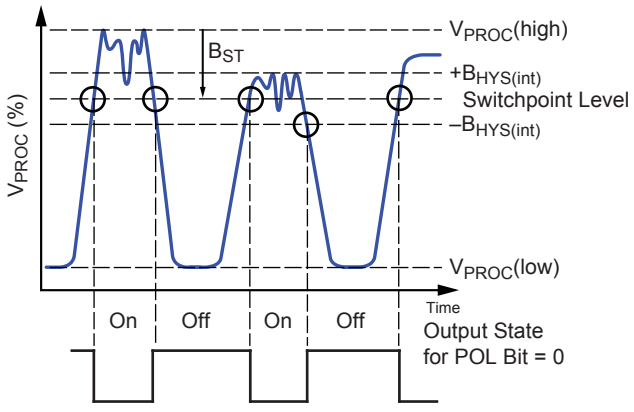


Figure 7. Switching on internal hysteresis

Operating Modes

Startup Hysteresis Mode

After power-on, a minimum amount of peak-to-peak magnetic movement is required before the ATS601 will begin generating output edges. This required signal movement threshold is referred to as the startup hysteresis. There are two programmable options for this startup hysteresis threshold. With narrow (N option) startup hysteresis, the device will switch to a farther air gap, but will have reduced immunity to magnetic overshoot, thus limiting its close air gap capability. With wide (W option) startup hysteresis, the device will have a reduced maximum air gap capability, but improved magnetic overshoot immunity. After the magnetic signal exceeds the startup hysteresis value for the first time, the device will transition to Calibration mode.

Calibration Mode

In Calibration mode, the ATS601 uses threshold based switching with continuous update. This ensures that all teeth and valleys are captured correctly, but provides slightly reduced accuracy relative to Running mode. The device stays in Calibration mode long enough to correctly capture enough peaks to fill the Running mode threshold memory. This corresponds to three output edges with the single-tooth update (1 option) and nine edges with the four-tooth update (4 option). After Calibration mode is complete, the device transitions to Running mode.

Running Mode

In Running mode the ATS601 uses threshold based switching with internal hysteresis described previously, in the Threshold Update and Switchpoints sections. The threshold update is intended to optimize output switching accuracy when used with common camshaft targets, including cases with runout and narrow target valleys.

Stop Mode

The ATS601 has an internal timer that begins counting on each output edge. If the timer reaches t_{SM} before another output edge occurs, the device moves into Stop mode. Stop mode is the same as Startup Hysteresis mode, but with a hysteresis value determined as a percentage of the previously measured target amplitude. Stop mode ensures no missed or extra output edges, even during situations with large temperature drifts and no target rotation.

Watchdog

The ATS601 has a peak detector continuously tracking the magnetic signal. If a sudden large signal change causes the sensor IC output to stop switching but the peak detector continues to detect valid signal movement, the watchdog will be fired. In case of a watchdog event, the sensor IC performs a self-reset and returns to the initial Startup Hysteresis mode to regain output switching.

Power Derating

Power Derating

The device must be operated below the maximum junction temperature of the device, T_J(max). Under certain combinations of peak conditions, reliable operation may require derating supplied power or improving the heat dissipation properties of the application. This section presents a procedure for correlating factors affecting operating T_J. (Thermal data is also available on the Allegro MicroSystems website.)

The Package Thermal Resistance, $R_{\theta JA}$, is a figure of merit summarizing the ability of the application and the device to dissipate heat from the junction (die), through all paths to the ambient air. Its primary component is the Effective Thermal Conductivity, K, of the printed circuit board, including adjacent devices and traces. Radiation from the die through the device case, $R_{\theta JC}$, is a relatively small component of $R_{\theta JA}$. Ambient air temperature, T_A, and air motion are significant external factors, damped by overmolding.

The effect of varying power levels (Power Dissipation, P_D), can be estimated. The following formulas represent the fundamental relationships used to estimate T_J, at P_D.

$$P_D = V_{IN} \times I_{IN} \quad (1)$$

$$\Delta T = P_D \times R_{\theta JA} \quad (2)$$

$$T_J = T_A + \Delta T \quad (3)$$

For example, given common conditions such as: T_A= 25°C, V_{CC} = 12 V, I_{CC} = 7 mA, and $R_{\theta JA}$ = 126 °C/W, then:

$$P_D = V_{CC} \times I_{CC} = 12\ V \times 7\ mA = 84\ mW$$

$$\Delta T = P_D \times R_{\theta JA} = 84\ mW \times 126\ °C/W = 10.6°C$$

$$T_J = T_A + \Delta T = 25°C + 10.6°C = 35.6°C$$

A worst-case estimate, P_D(max), represents the maximum allowable power level (V_{CC}(max), I_{CC}(max)), without exceeding T_J(max), at a selected $R_{\theta JA}$ and T_A.

Example: Reliability for V_{CC} at T_A=150°C, package SG, using single layer PCB.

Observe the worst-case ratings for the device, specifically: $R_{\theta JA}$=126°C/W, T_J(max)=165°C, $V_{CC(absmax)}$=24 V, and I_{CC} = 10 mA.

Calculate the maximum allowable power level, P_D(max). First, invert equation 3:

$$\Delta T(max) = T_J(max) - T_A = 165\,°C - 150\,°C = 15\,°C$$

This provides the allowable increase to T_J resulting from internal power dissipation. Then, invert equation 2:

$$P_D(max) = \Delta T(max) \div R_{\theta JA} = 15°C \div 126\ °C/W = 119\,mW$$

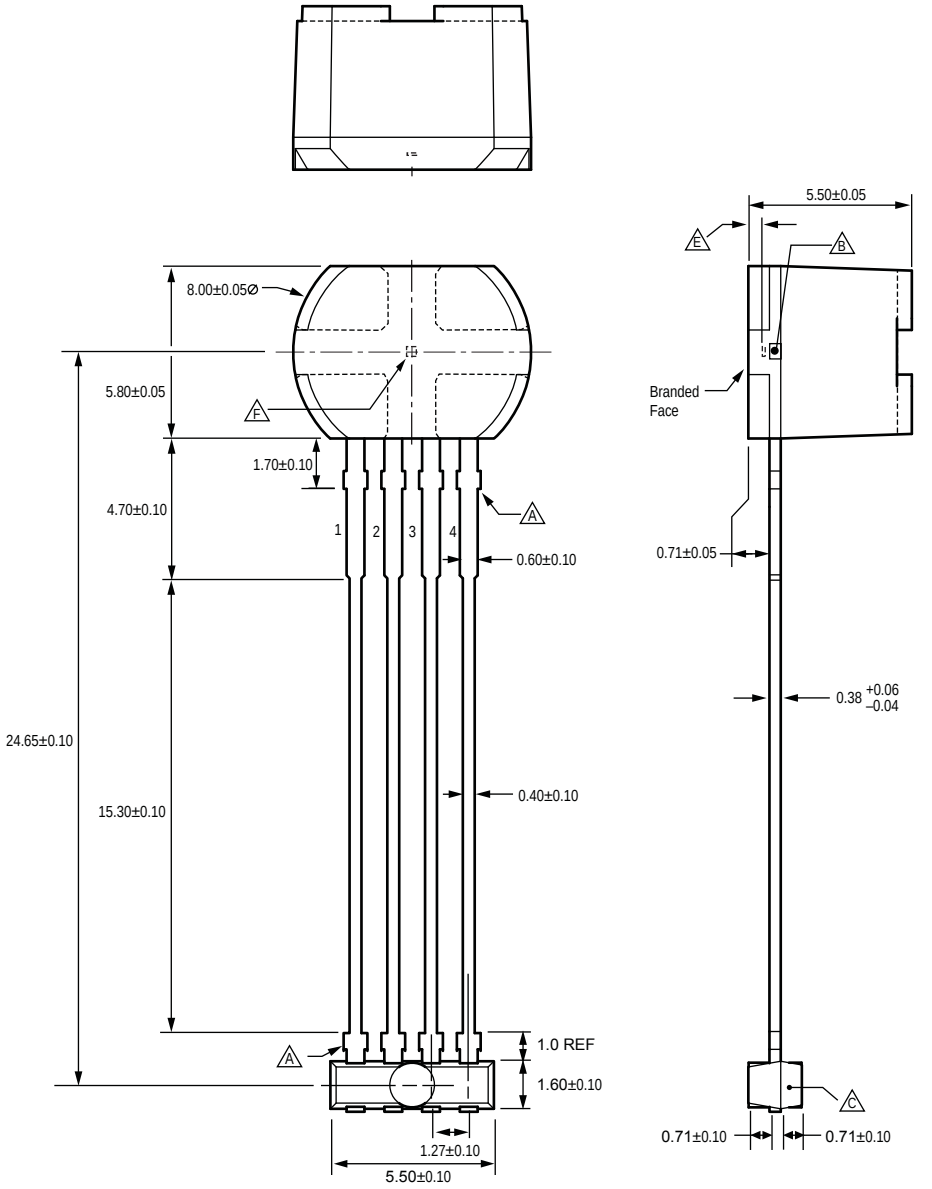
Finally, invert equation 1 with respect to voltage:

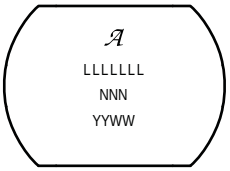
$$V_{CC(est)} = P_D(max) \div I_{CC} = 119\,mW \div 10\,mA = 11.9\ V$$

The result indicates that, at T_A, the application and device can dissipate adequate amounts of heat at voltages ≤$V_{CC(est)}$.

Compare $V_{CC(est)}$ to V_{CC}(max). If $V_{CC(est)} \leq V_{CC}$(max), then reliable operation between $V_{CC(est)}$ and V_{CC}(max) requires enhanced $R_{\theta JA}$. If $V_{CC(est)} \geq V_{CC(max)}$, then operation between $V_{CC(est)}$ and V_{CC}(max) is reliable under these conditions.

Package SG, 4-Pin SIP





D Standard Branding Reference View

A = Supplier emblem
L = Lot identifier
N = Last three numbers of device part number
Y = Last two digits of year of manufacture
W = Week of manufacture

For Reference Only, not for tooling use (reference DWG-9002)
Dimensions in millimeters

A Dambar removal protrusion (16X)

B Metallic protrusion, electrically connected to pin 4 and substrate (both sides)

C Thermoplastic Molded Lead Bar for alignment during shipment

D Branding scale and appearance at supplier discretion

E Active Area Depth, 0.43 mm

F Hall element, not to scale

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