MC74HC4060A

14-Stage Binary Ripple Counter With Oscillator

High−Performance Silicon−Gate CMOS

The MC74HC4060A is identical in pinout to the standard CMOS MC14060B. The device inputs are compatible with standard CMOS outputs; with pullup resistors, they are compatible with LSTTL outputs.

This device consists of 14 master−slave flip−flops and an oscillator with a frequency that is controlled either by a crystal or by an RC circuit connected externally. The output of each flip−flop feeds the next and the frequency at each output is half of that of the preceding one. The state of the counter advances on the negative−going edge of the Osc In. The active−high Reset is asynchronous and disables the oscillator to allow very low power consumption during stand−by operation.

State changes of the Q outputs do not occur simultaneously because of internal ripple delays. Therefore, decoded output signals are subject to decoding spikes and may have to be gated with Osc Out 2 of the HC4060A.

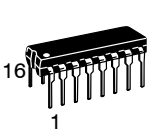
Features

- Output Drive Capability: 10 LSTTL Loads
- Outputs Directly Interface to CMOS, NMOS, and TTL
- Operating Voltage Range: 2.0 to 6.0 V
- Low Input Current: 1 μA
- High Noise Immunity Characteristic of CMOS Devices
- In Compliance With JEDEC Standard No. 7A Requirements
- Chip Complexity: 390 FETs or 97.5 Equivalent Gates
- NLV Prefix for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC−Q100 Qualified and PPAP Capable
- These Devices are Pb−Free, Halogen Free/BFR Free and are RoHS Compliant

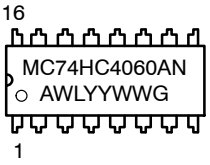
ON Semiconductor®

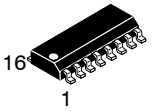
http://onsemi.com

MARKING DIAGRAMS

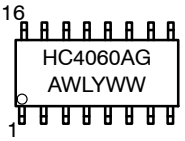


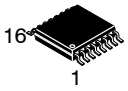
PDIP−16
N SUFFIX
CASE 648



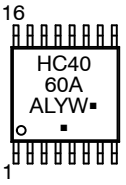


SOIC−16
D SUFFIX
CASE 751B





TSSOP−16
DT SUFFIX
CASE 948F



A	=	Assembly Location
L, WL	=	Wafer Lot
Y, YY	=	Year
W, WW	=	Work Week
G or ▪	=	Pb−Free Package

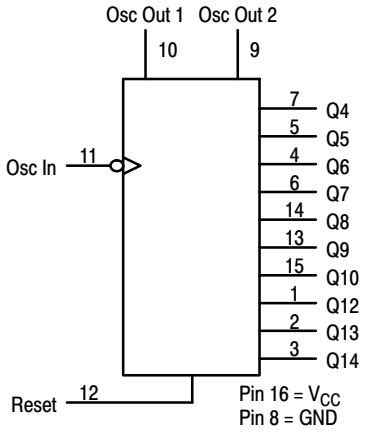
(Note: Microdot may be in either location)

ORDERING INFORMATION

See detailed ordering and shipping information in the package dimensions section on page 2 of this data sheet.

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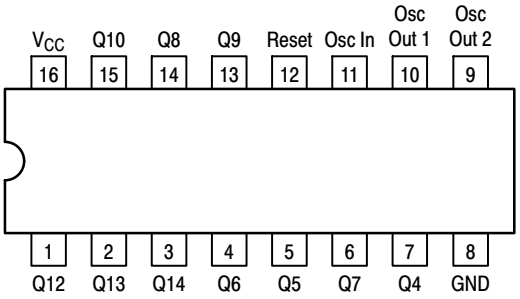
LOGIC DIAGRAM



FUNCTION TABLE

Clock	Reset	Output State
⌐‾	L	No Change
‾⌐	L	Advance to Next State
X	H	All Outputs Are Low

Pinout: 16−Lead Plastic Package (Top View)



ORDERING INFORMATION

Device	Package	Shipping†
MC74HC4060ANG	PDIP−16 (Pb−Free)	500 Units / Box
MC74HC4060ADG	SOIC−16 (Pb−Free)	48 Units / Rail
MC74HC4060ADR2G		2500 Units / Reel
NLV74HC4060ADR2G*		2500 Units / Reel
MC74HC4060ADTG	TSSOP−16 (Pb−Free)	96 Units / Rail
MC74HC4060ADTR2G		2500 Units / Reel
NLVHC4060ADTR2G*		2500 Units / Reel

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

*NLV Prefix for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC−Q100 Qualified and PPAP Capable.

MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V_{CC}	DC Supply Voltage (Referenced to GND)	– 0.5 to + 7.0	V
V_{in}	DC Input Voltage (Referenced to GND)	– 0.5 to V_{CC} + 0.5	V
V_{out}	DC Output Voltage (Referenced to GND)	– 0.5 to V_{CC} + 0.5	V
I_{in}	DC Input Current, per Pin	± 20	mA
I_{out}	DC Output Current, per Pin	± 25	mA
I_{CC}	DC Supply Current, V_{CC} and GND Pins	± 50	mA
P_D	Power Dissipation in Still Air, Plastic DIP† SOIC Package† TSSOP Package†	750 500 450	mW
T_{stg}	Storage Temperature Range	– 65 to + 150	°C
T_L	Lead Temperature, 1 mm from Case for 10 Seconds Plastic DIP, SOIC or TSSOP Package	260	°C

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

†Derating — Plastic DIP: – 10 mW/°C from 65° to 125°C
SOIC Package: – 7 mW/°C from 65° to 125°C
TSSOP Package: – 6.1 mW/°C from 65° to 125°C

This device contains protection circuitry to guard against damage due to high static voltages or electric fields. However, precautions must be taken to avoid applications of any voltage higher than maximum rated voltages to this high–impedance circuit. For proper operation, V_{in} and V_{out} should be constrained to the range GND ≤ (V_{in} or V_{out}) ≤ V_{CC}.

Unused inputs must always be tied to an appropriate logic voltage level (e.g., either GND or V_{CC}). Unused outputs must be left open.

RECOMMENDED OPERATING CONDITIONS

Symbol	Parameter		Min	Max	Unit
V_{CC}	DC Supply Voltage (Referenced to GND)		2.5*	6.0	V
V_{in}, V_{out}	DC Input Voltage, Output Voltage (Referenced to GND)		0	V_{CC}	V
T_A	Operating Temperature Range, All Package Types		– 55	+ 125	°C
t_r, t_f	Input Rise/Fall Time (Figure 1)	V_{CC} = 2.0 V V_{CC} = 4.5 V V_{CC} = 6.0 V	0 0 0	1000 500 400	ns

*The oscillator is guaranteed to function at 2.5 V minimum. However, parametrics are tested at 2.0 V by driving Pin 11 with an external clock source.

DC CHARACTERISTICS (Voltages Referenced to GND)

Symbol	Parameter	Condition	V_{CC} V	Guaranteed Limit –55 to 25°C	Guaranteed Limit ≤85°C	Guaranteed Limit ≤125°C	Unit
V_{IH}	Minimum High–Level Input Voltage	V_{out} = 0.1V or V_{CC} –0.1V $\|I_{out}\| \le 20\mu A$	2.0 3.0 4.5 6.0	1.50 2.10 3.15 4.20	1.50 2.10 3.15 4.20	1.50 2.10 3.15 4.20	V
V_{IL}	Maximum Low–Level Input Voltage	V_{out} = 0.1V or V_{CC} – 0.1V $\|I_{out}\| \le 20\mu A$	2.0 3.0 4.5 6.0	0.50 0.90 1.35 1.80	0.50 0.90 1.35 1.80	0.50 0.90 1.35 1.80	V
V_{OH}	Minimum High–Level Output Voltage (Q4–Q10, Q12–Q14)	V_{in} = V_{IH} or V_{IL} $\|I_{out}\| \le 20\mu A$	2.0 4.5 6.0	1.9 4.4 5.9	1.9 4.4 5.9	1.9 4.4 5.9	V
		V_{in} =V_{IH} or V_{IL} $\|I_{out}\| \le 2.4mA$ $\|I_{out}\| \le 4.0mA$ $\|I_{out}\| \le 5.2mA$	3.0 4.5 6.0	2.48 3.98 5.48	2.34 3.84 5.34	2.20 3.70 5.20	

DC CHARACTERISTICS (Voltages Referenced to GND)

Symbol	Parameter	Condition	V_{CC} V	Guaranteed Limit −55 to 25°C	Guaranteed Limit ≤85°C	Guaranteed Limit ≤125°C	Unit
V_{OL}	Maximum Low−Level Output Voltage (Q4−Q10, Q12−Q14)	$V_{in} = V_{IH}$ or V_{IL} $\|I_{out}\| \le 20\mu A$	2.0 4.5 6.0	0.1 0.1 0.1	0.1 0.1 0.1	0.1 0.1 0.1	V
		$V_{in} = V_{IH}$ or V_{IL} $\|I_{out}\| \le 2.4mA$ $\|I_{out}\| \le 4.0mA$ $\|I_{out}\| \le 5.2mA$	3.0 4.5 6.0	0.26 0.26 0.26	0.33 0.33 0.33	0.40 0.40 0.40	
V_{OH}	Minimum High−Level Output Voltage (Osc Out 1, Osc Out 2)	$V_{in} = V_{CC}$ or GND $\|I_{out}\| \le 20\mu A$	2.0 4.5 6.0	1.9 4.4 5.9	1.9 4.4 5.9	1.9 4.4 5.9	V
		$V_{in} = V_{CC}$ or GND $\|I_{out}\| \le 0.7mA$ $\|I_{out}\| \le 1.0mA$ $\|I_{out}\| \le 1.3mA$	3.0 4.5 6.0	2.48 3.98 5.48	2.34 3.84 5.34	2.20 3.70 5.20	
V_{OL}	Maximum Low−Level Output Voltage (Osc Out 1, Osc Out 2)	$V_{in} = V_{CC}$ or GND $\|I_{out}\| \le 20\mu A$	2.0 4.5 6.0	0.1 0.1 0.1	0.1 0.1 0.1	0.1 0.1 0.1	V
		$V_{in} = V_{CC}$ or GND $\|I_{out}\| \le 0.7mA$ $\|I_{out}\| \le 1.0mA$ $\|I_{out}\| \le 1.3mA$	3.0 4.5 6.0	0.26 0.26 0.26	0.33 0.33 0.33	0.40 0.40 0.40	
I_{in}	Maximum Input Leakage Current	$V_{in} = V_{CC}$ or GND	6.0	±0.1	±1.0	±1.0	μA
I_{CC}	Maximum Quiescent Supply Current (per Package)	$V_{in} = V_{CC}$ or GND $I_{out} = 0\mu A$	6.0	4	40	160	μA

AC CHARACTERISTICS (C_L = 50 pF, Input $t_r = t_f$ = 6 ns)

Symbol	Parameter	V_{CC} V	Guaranteed Limit −55 to 25°C	Guaranteed Limit ≤85°C	Guaranteed Limit ≤125°C	Unit
f_{max}	Maximum Clock Frequency (50% Duty Cycle) (Figures 1 and 4)	2.0 3.0 4.5 6.0	6.0 10 30 50	9.0 14 28 45	8.0 12 25 40	MHz
t_{PLH}, t_{PHL}	Maximum Propagation Delay, Osc In to Q4* (Figures 1 and 4)	2.0 3.0 4.5 6.0	300 180 60 51	375 200 75 64	450 250 90 75	ns
t_{PLH}, t_{PHL}	Maximum Propagation Delay, Osc In to Q14* (Figures 1 and 4)	2.0 3.0 4.5 6.0	500 350 250 200	750 450 275 220	1000 600 300 250	ns
t_{PHL}	Maximum Propagation Delay, Reset to Any Q (Figures 2 and 4)	2.0 3.0 4.5 6.0	195 75 39 33	245 100 49 42	300 125 61 53	ns
t_{PLH}, t_{PHL}	Maximum Propagation Delay, Qn to Qn+1 (Figures 3 and 4)	2.0 3.0 4.5 6.0	75 60 15 13	95 75 19 16	125 95 24 20	ns

AC CHARACTERISTICS (C_L = 50 pF, Input $t_r = t_f$ = 6 ns) **– continued**

Symbol	Parameter	V_{CC} V	Guaranteed Limit			Unit
			–55 to 25°C	≤85°C	≤125°C	
t_{TLH}, t_{THL}	Maximum Output Transition Time, Any Output (Figures 1 and 4)	2.0	75	95	110	ns
		3.0	27	32	36	
		4.5	15	19	22	
		6.0	13	16	19	
C_{in}	Maximum Input Capacitance		10	10	10	pF

* For T_A = 25°C and C_L = 50 pF, typical propagation delay from Clock to other Q outputs may be calculated with the following equations:
V_{CC} = 2.0 V: t_P = [93.7 + 59.3 (n–1)] ns V_{CC} = 4.5 V: t_P = [30.25 + 14.6 (n–1)] ns
V_{CC} = 3.0 V: t_P = [61.5+ 34.4 (n–1)] ns V_{CC} = 6.0 V: t_P = [24.4 + 12 (n–1)] ns

		Typical @ 25°C, V_{CC} = 5.0 V	
C_{PD}	Power Dissipation Capacitance (Per Package)*	35	pF

* Used to determine the no–load dynamic power consumption: $P_D = C_{PD} V_{CC}^2 f + I_{CC} V_{CC}$.

TIMING REQUIREMENTS (Input $t_r = t_f$ = 6 ns)

Symbol	Parameter	V_{CC} V	Guaranteed Limit			Unit
			–55 to 25°C	≤85°C	≤125°C	
t_{rec}	Minimum Recovery Time, Reset Inactive to Clock (Figure 2)	2.0	100	125	150	ns
		3.0	75	100	120	
		4.5	20	25	30	
		6.0	17	21	25	
t_w	Minimum Pulse Width, Clock (Figure 1)	2.0	75	95	110	ns
		3.0	27	32	36	
		4.5	15	19	23	
		6.0	13	16	19	
t_w	Minimum Pulse Width, Reset (Figure 2)	2.0	75	95	110	ns
		3.0	27	32	36	
		4.5	15	19	23	
		6.0	13	16	19	
t_r, t_f	Maximum Input Rise and Fall Times (Figure 1)	2.0	1000	1000	1000	ns
		3.0	800	800	800	
		4.5	500	500	500	
		6.0	400	400	400	

PIN DESCRIPTIONS

INPUTS

Osc In (Pin 11)

Negative−edge triggering clock input. A high−to−low transition on this input advances the state of the counter. Osc In may be driven by an external clock source.

Reset (Pin 12)

Active−high reset. A high level applied to this input asynchronously resets the counter to its zero state (forcing all Q outputs low) and disables the oscillator.

OUTPUTS

Q4—Q10, Q12−Q14 (Pins 7, 5, 4, 6, 13, 15, 1, 2, 3)

Active−high outputs. Each Qn output divides the Clock input frequency by 2^N. The user should note the Q1, Q2, Q3 and Q11 are not available as outputs.

Osc Out 1, Osc Out 2 (Pins 9, 10)

Oscillator outputs. These pins are used in conjunction with Osc In and the external components to form an oscillator. When Osc In is being driven with an external clock source, Osc Out 1 and Osc Out 2 must be left open circuited. With the crystal oscillator configuration in Figure 6, Osc Out 2 must be left open circuited.

SWITCHING WAVEFORMS

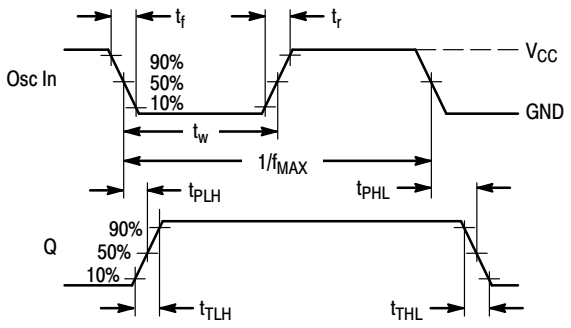


Figure 1.

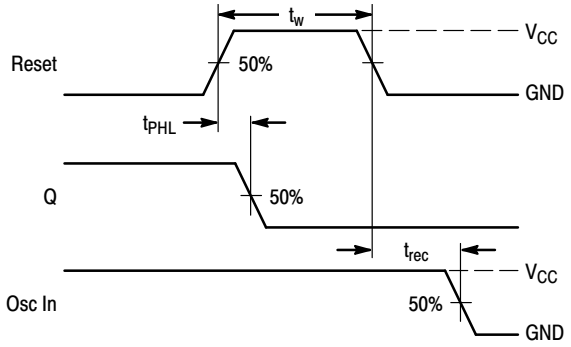


Figure 2.

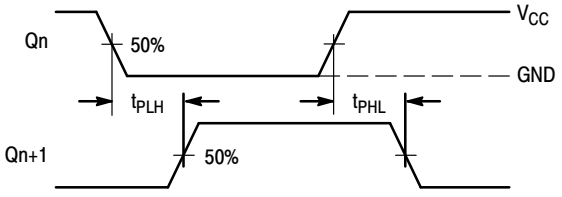
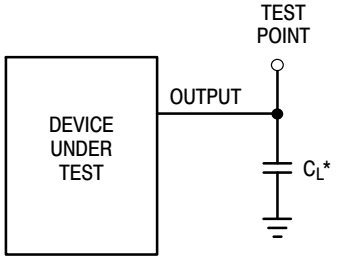


Figure 3.



*Includes all probe and jig capacitance

Figure 4. Test Circuit

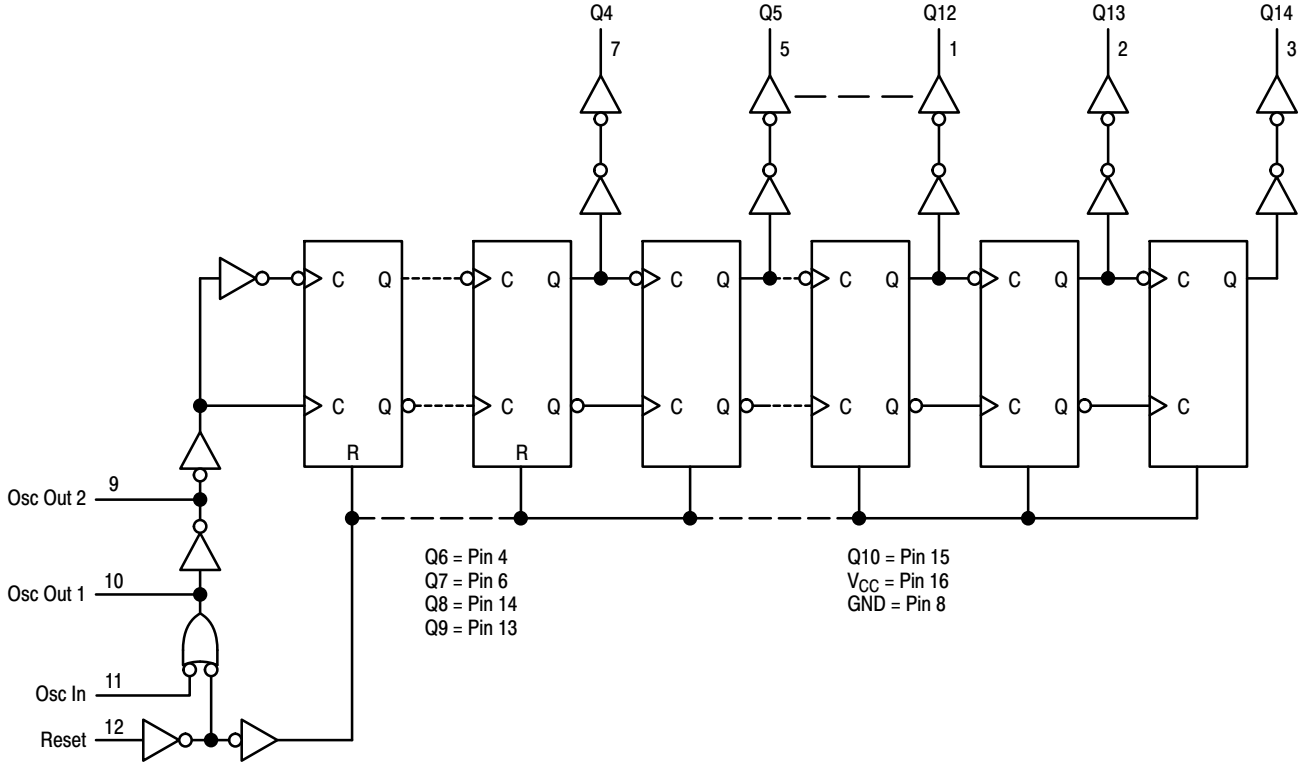


Figure 5. Expanded Logic Diagram

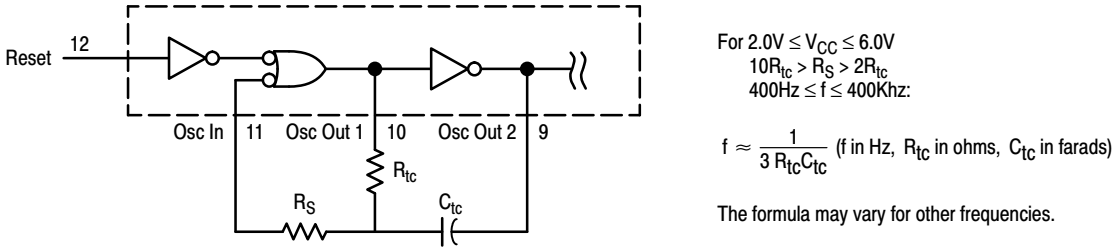


Figure 6. Oscillator Circuit Using RC Configuration

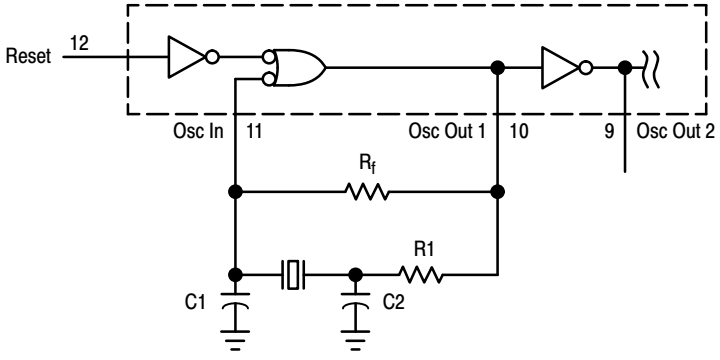
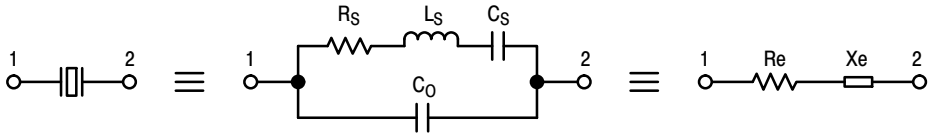


Figure 7. Pierce Crystal Oscillator Circuit

TABLE 1. CRYSTAL OSCILLATOR AMPLIFIER SPECIFICATIONS (T_A = 25°C; Input = Pin 11, Output = Pin 10)

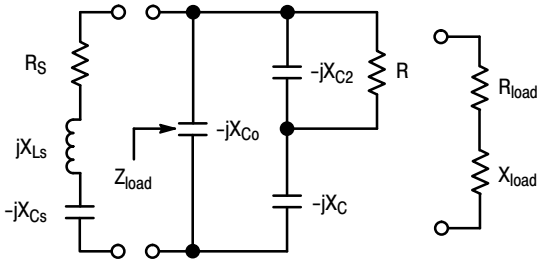
Type		Positive Reactance (Pierce)
Input Resistance, R_{in}		60MΩ Minimum
Output Impedance, Z_{out} (4.5V Supply)		200Ω (See Text)
Input Capacitance, C_{in}		5pF Typical
Output Capacitance, C_{out}		7pF Typical
Series Capacitance, C_a		5pF Typical
Open Loop Voltage Gain with Output at Full Swing, α	3Vdc Supply 4Vdc Supply 5Vdc Supply 6Vdc Supply	5.0 Expected Minimum 4.0 Expected Minimum 3.3 Expected Minimum 3.1 Expected Minimum

PIERCE CRYSTAL OSCILLATOR DESIGN



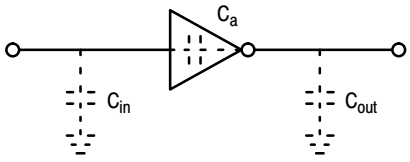
Value are supplied by crystal manufacturer (parallel resonant crystal).

Figure 8. Equivalent Crystal Networks



NOTE: C = C1 + C_{in} and R = R1 + R_{out}. C_o is considered as part of the load. C_a and R_f typically have minimal effect below 2MHz.

Figure 9. Series Equivalent Crystal Load



Values are listed in Table 1.

Figure 10. Parasitic Capacitances of the Amplifier

DESIGN PROCEDURES

The following procedure applies for oscillators operating below 2MHz where Z is a resistor R1. Above 2MHz, additional impedance elements should be considered: C_{out} and C_a of the amp, feedback resistor R_f, and amplifier phase shift error from 180°C.

Step 1: Calculate the equivalent series circuit of the crystal at the frequency of oscillation.

$$Z_e = \frac{-jX_{C_o}(R_s + jX_{L_s} - jX_{C_s})}{-jX_{C_o} + R_s + jX_{L_s} - jX_{C_s}} = R_e + jX_e$$

Reactance jX_e should be positive, indicating that the crystal is operating as an inductive reactance at the oscillation frequency. The maximum R_s for the crystal should be used in the equation.

Step 2: Determine β, the attenuation, of the feedback network. For a closed-loop gain of 2,$A_v\beta = 2,\beta = 2/A_v$ where A_v is the gain of the HC4060A amplifier.

Step 3: Determine the manufacturer's loading capacitance. For example: A manufacturer may specify an external load capacitance of 32pF at the required frequency.

Step 4: Determine the required Q of the system, and calculate R_{load}, For example, a manufacturer specifies a crystal Q of 100,000. In-circuit Q is arbitrarily set at 20% below crystal Q or 80,000. Then $R_{load} = (2\pi f_o L_S/Q) - R_s$ where L_s and R_s are crystal parameters.

Step 5: Simultaneously solve, using a computer,

$$\beta = \frac{X_C \cdot X_{C2}}{R \cdot R_e + X_{C2}(X_e - X_C)} \quad \text{(with feedback phase shift = 180°)} \qquad (\text{Eq 1})$$

$$X_e = X_{C2} + X_C + \frac{R_e X_{C2}}{R} = X_{C_{load}} \quad \text{(where the loading capacitor is an external load, not including } C_o) \qquad (\text{Eq 2})$$

$$R_{load} = \frac{RX_{C_o}X_{C2}[(X_C + X_{C2})(X_C + X_{C_o}) - X_C(X_C + X_{C_o} + X_{C2})]}{X^2_{C2}(X_C + X_{C_o})^2 + R^2(X_C + X_{C_o} + X_{C2})^2} \qquad (\text{Eq 3})$$

Here $R = R_{out} + R1$. R_{out} is amp output resistance, R1 is Z. The C corresponding to X_C is given by $C = C1 + C_{in}$.

Alternately, pick a value for R1 (i.e, let $R1 = R_S$). Solve Equations 1 and 2 for C1 and C2. Use Equation 3 and the fact that $Q = 2\pi f_o L_s/(R_s + R_{load})$ to find in-circuit Q. If Q is not satisfactory pick another value for R1 and repeat the procedure.

CHOOSING R1

Power is dissipated in the effective series resistance of the crystal. The drive level specified by the crystal manufacturer is the maximum stress that a crystal can withstand without damage or excessive shift in frequency. R1 limits the drive level.

To verify that the maximum dc supply voltage does not overdrive the crystal, monitor the output frequency as a function of voltage at Osc Out 2 (Pin 9). The frequency should increase very slightly as the dc supply voltage is increased. An overdriven crystal will decrease in frequency or become unstable with an increase in supply voltage. The operating supply voltage must be reduced or R1 must be increased in value if the overdriven condition exists. The user should note that the oscillator start-up time is proportional to the value of R1.

SELECTING R_f

The feedback resistor, R_f, typically ranges up to 20MΩ. R_f determines the gain and bandwidth of the amplifier. Proper bandwidth insures oscillation at the correct frequency plus roll-off to minimize gain at undesirable frequencies, such as the first overtone. R_f must be large enough so as to not affect the phase of the feedback network in an appreciable manner.

ACKNOWLEDGEMENTS AND RECOMMENDED REFERENCES

The following publications were used in preparing this data sheet and are hereby acknowledged and recommended for reading:

Technical Note TN-24, Statek Corp.

Technical Note TN-7, Statek Corp.

D. Babin, "Designing Crystal Oscillators", Machine Design, March 7, 1985.

D. Babin, "Guidelines for Crystal Oscillator Design", Machine Design, April 25, 1985.

ALSO RECOMMENDED FOR READING:

E. Hafner, "The Piezoelectric Crystal Unit-Definitions and Method of Measurement", Proc. IEEE, Vol. 57, No. 2, Feb., 1969.

D. Kemper, L. Rosine, "Quartz Crystals for Frequency Control", Electro-Technology, June, 1969.

P. J. Ottowitz, "A Guide to Crystal Selection", Electronic Design, May, 1966.

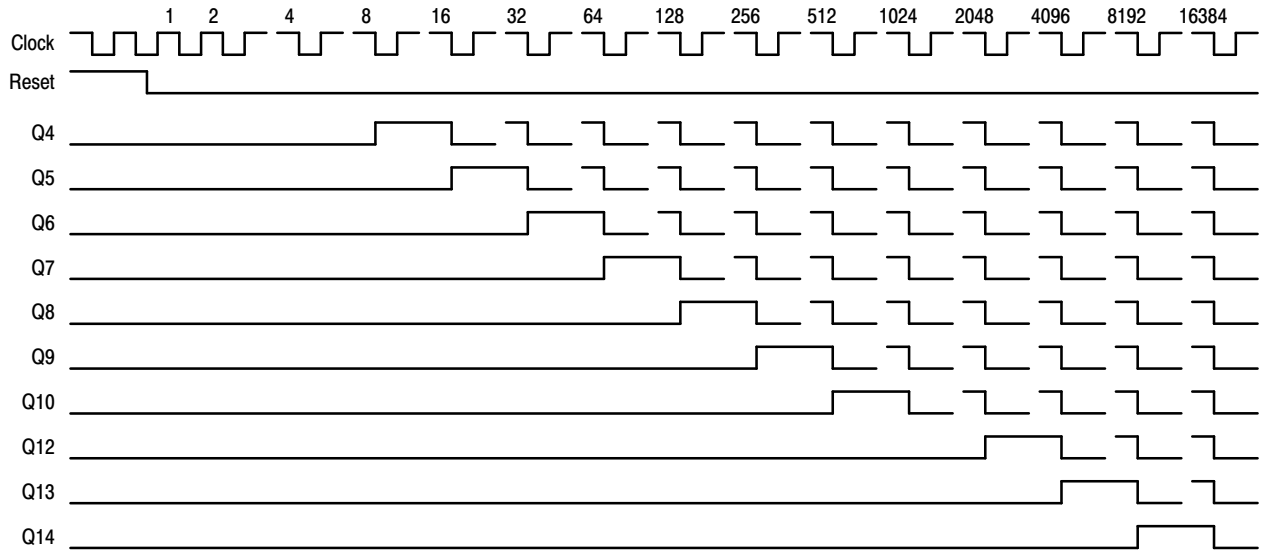
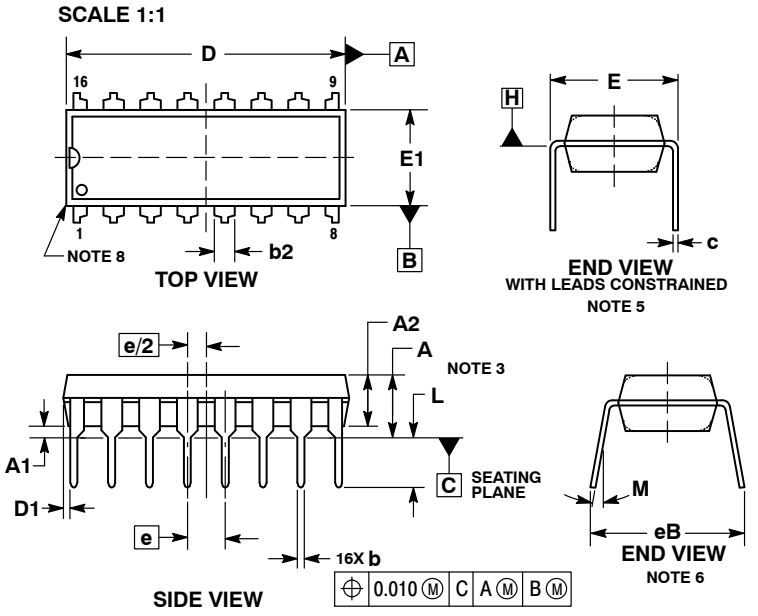


Figure 11. Timing Diagram

PACKAGE DIMENSIONS

PDIP−16
CASE 648−08
ISSUE U

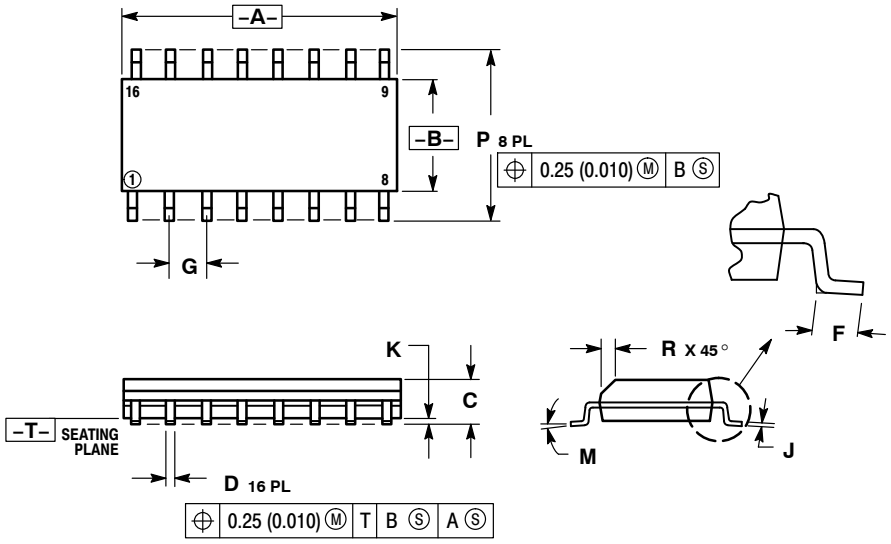


NOTES:
1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
2. CONTROLLING DIMENSION: INCHES.
3. DIMENSIONS A, A1 AND L ARE MEASURED WITH THE PACKAGE SEATED IN JEDEC SEATING PLANE GAUGE GS−3.
4. DIMENSIONS D, D1 AND E1 DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS. MOLD FLASH OR PROTRUSIONS ARE NOT TO EXCEED 0.10 INCH.
5. DIMENSION E IS MEASURED AT A POINT 0.015 BELOW DATUM PLANE H WITH THE LEADS CONSTRAINED PERPENDICULAR TO DATUM C.
6. DIMENSION E3 IS MEASURED AT THE LEAD TIPS WITH THE LEADS UNCONSTRAINED.
7. DATUM PLANE H IS COINCIDENT WITH THE BOTTOM OF THE LEADS, WHERE THE LEADS EXIT THE BODY.
8. PACKAGE CONTOUR IS OPTIONAL (ROUNDED OR SQUARE CORNERS).

	INCHES		MILLIMETERS	
DIM	MIN	MAX	MIN	MAX
A	----	0.210	---	5.33
A1	0.015	----	0.38	---
A2	0.115	0.195	2.92	4.95
b	0.014	0.022	0.35	0.56
b2	0.060 TYP		1.52 TYP	
C	0.008	0.014	0.20	0.36
D	0.735	0.775	18.67	19.69
D1	0.005	----	0.13	---
E	0.300	0.325	7.62	8.26
E1	0.240	0.280	6.10	7.11
e	0.100 BSC		2.54 BSC	
eB	----	0.430	---	10.92
L	0.115	0.150	2.92	3.81
M	----	10°	---	10°

PACKAGE DIMENSIONS

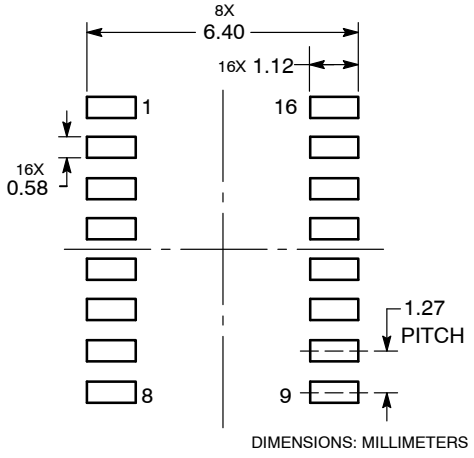
SOIC−16
CASE 751B−05
ISSUE K



NOTES:
1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: MILLIMETER.
3. DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION.
4. MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
5. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.

	MILLIMETERS		INCHES	
DIM	MIN	MAX	MIN	MAX
A	9.80	10.00	0.386	0.393
B	3.80	4.00	0.150	0.157
C	1.35	1.75	0.054	0.068
D	0.35	0.49	0.014	0.019
F	0.40	1.25	0.016	0.049
G	1.27 BSC		0.050 BSC	
J	0.19	0.25	0.008	0.009
K	0.10	0.25	0.004	0.009
M	0°	7°	0°	7°
P	5.80	6.20	0.229	0.244
R	0.25	0.50	0.010	0.019

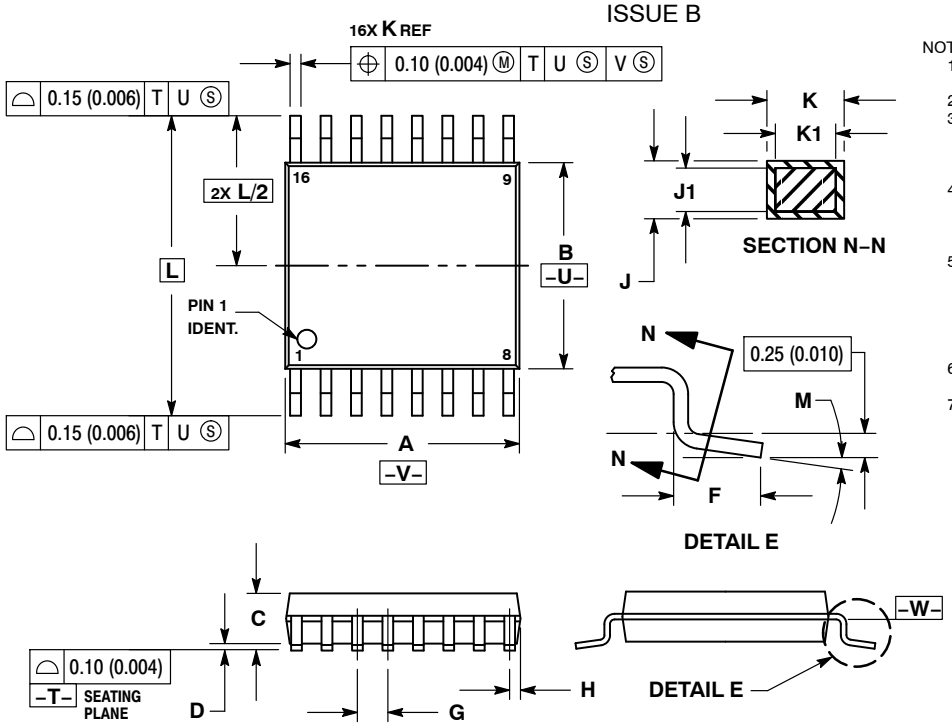
SOLDERING FOOTPRINT



DIMENSIONS: MILLIMETERS

PACKAGE DIMENSIONS

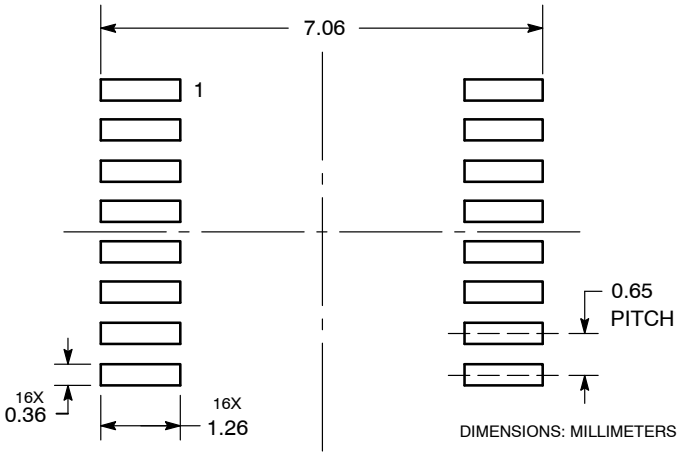
TSSOP−16
CASE 948F
ISSUE B



NOTES:
1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: MILLIMETER.
3. DIMENSION A DOES NOT INCLUDE MOLD FLASH. PROTRUSIONS OR GATE BURRS. MOLD FLASH OR GATE BURRS SHALL NOT EXCEED 0.15 (0.006) PER SIDE.
4. DIMENSION B DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR PROTRUSION SHALL NOT EXCEED 0.25 (0.010) PER SIDE.
5. DIMENSION K DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.08 (0.003) TOTAL IN EXCESS OF THE K DIMENSION AT MAXIMUM MATERIAL CONDITION.
6. TERMINAL NUMBERS ARE SHOWN FOR REFERENCE ONLY.
7. DIMENSION A AND B ARE TO BE DETERMINED AT DATUM PLANE −W−.

	MILLIMETERS		INCHES	
DIM	MIN	MAX	MIN	MAX
A	4.90	5.10	0.193	0.200
B	4.30	4.50	0.169	0.177
C	−−−	1.20	−−−	0.047
D	0.05	0.15	0.002	0.006
F	0.50	0.75	0.020	0.030
G	0.65 BSC		0.026 BSC	
H	0.18	0.28	0.007	0.011
J	0.09	0.20	0.004	0.008
J1	0.09	0.16	0.004	0.006
K	0.19	0.30	0.007	0.012
K1	0.19	0.25	0.007	0.010
L	6.40 BSC		0.252 BSC	
M	0 °	8 °	0 °	8 °

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